

# Modifying a Self-Sensing Circuit to Increase the Stability of Vibration Control

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# Piezoelectric materials (PZTs) have properties that make it attractive as a sensor.

## Advantages

Non-intrusive

Has potential for self-diagnostic capabilities

High Strain Sensitivity

## Challenges

PZT is brittle

High electric fields are required ( $.5\text{-}2\text{ MV/m}$ )

Only low strains are obtainable

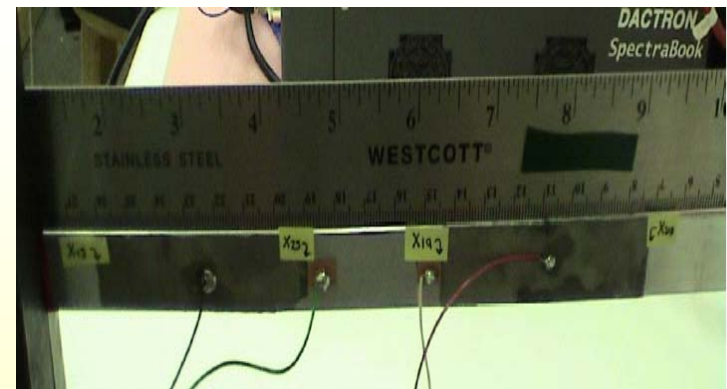
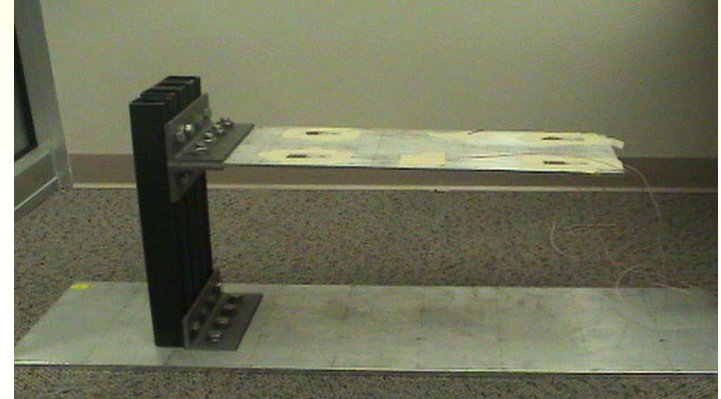


# Advantages of piezoelectric self-sensing actuators.

**Lighter and less costly than non-collocated systems.[Tani 2002]**

**The control force is applied where the response is measured.[Dosch 1992]**

**Unconditionally stable for feedback control.**

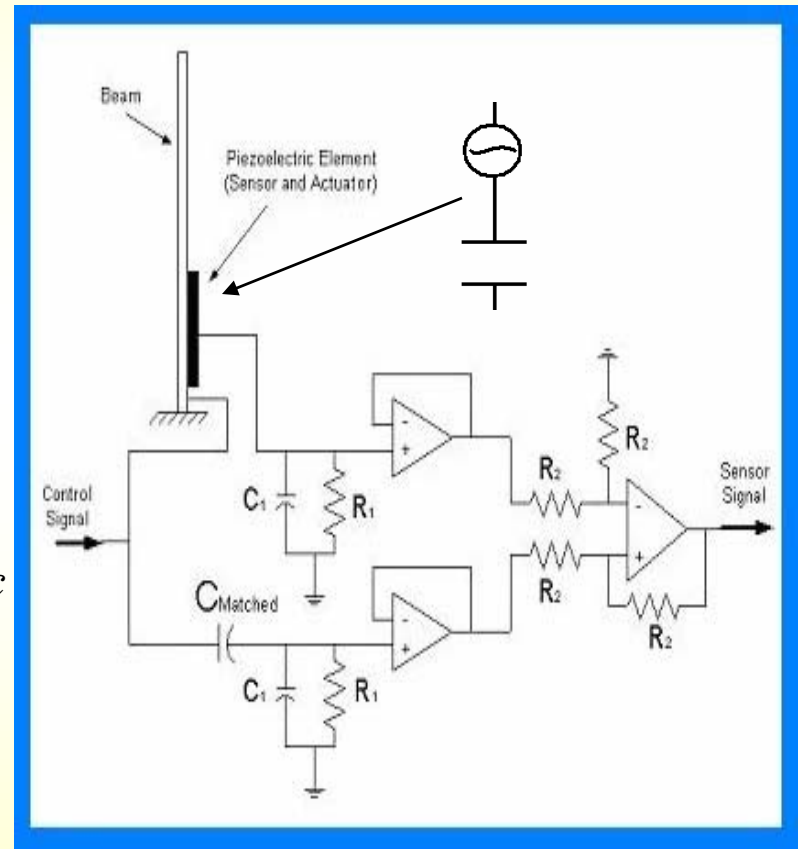


# Self-sensing actuators mix sensing and actuator voltages.

The control voltage is larger than the sensor signal.

Bridge circuits are used to separate the two signals.

$$V_s = \frac{C_p \cdot R \cdot s}{R(C_1 \cdot s + C_p \cdot s) + 1} V_c - \frac{C_m \cdot R \cdot s}{R(C_1 \cdot s + C_m \cdot s) + 1} V_c + \frac{C_p \cdot R \cdot s}{R(C_1 \cdot s + C_p \cdot s) + 1} V_p$$



Sodano 2003

# Challenge: Bridge circuits that distinguish the control and sensor signals are easily unbalanced.

**C<sub>p</sub> is temperature dependant.**

**Control instability results from the unbalanced circuit.**

**Tremendous research efforts have been dedicated to increase stability of the SS actuation.**

**Our goals:**

**Understand dynamic characteristics**

**Increase robustness of bridge circuit**

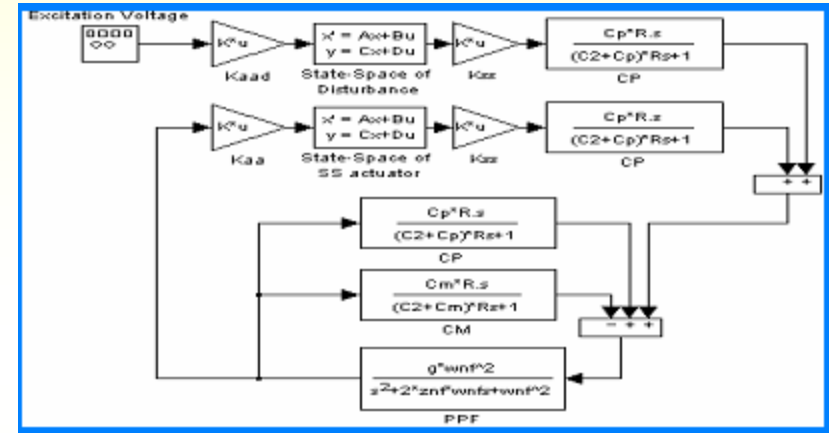
$$V_s = \frac{C_p \cdot R \cdot s}{R(C_1 \cdot s + C_p \cdot s) + 1} V_c - \frac{C_m \cdot R \cdot s}{R(C_1 \cdot s + C_m \cdot s) + 1} V_c + \frac{C_p \cdot R \cdot s}{R(C_1 \cdot s + C_p \cdot s) + 1} V_p$$

# This talk will cover modifications made to the bridge circuit to increase stability of the controller.

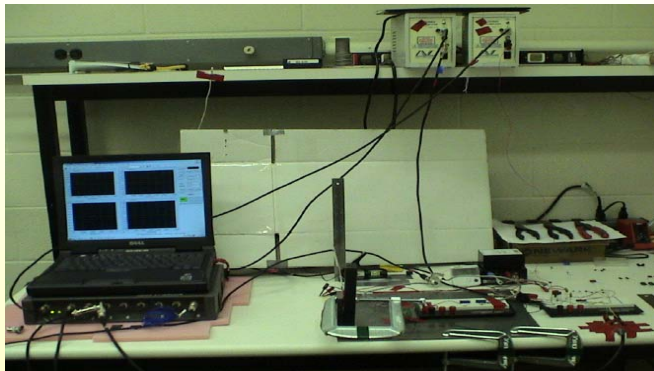
## 1. Analytical Modeling

$$A = \begin{bmatrix} 0 & 0 & 0 & 0 & 1 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 1 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 0 & 1 \\ -\omega_1^2 & 0 & 0 & 0 & -2\zeta\omega_1 & 0 & 0 & 0 \\ 0 & -\omega_2^2 & 0 & 0 & 0 & -2\zeta\omega_2 & 0 & 0 \\ 0 & 0 & -\omega_3^2 & 0 & 0 & 0 & -2\zeta\omega_3 & 0 \\ 0 & 0 & 0 & -\omega_4^2 & 0 & 0 & 0 & -2\zeta\omega_4 \end{bmatrix}$$

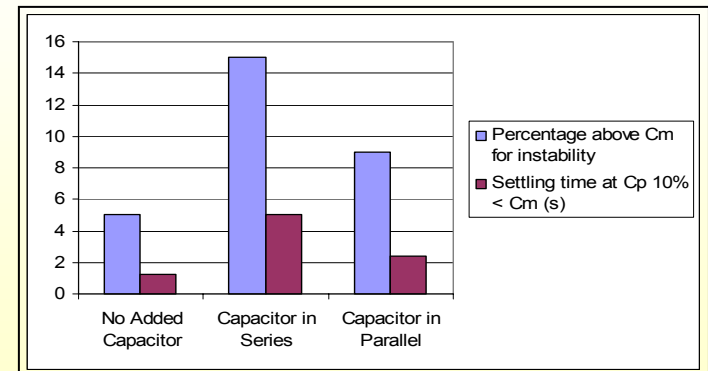
## 2. Analytical Simulation



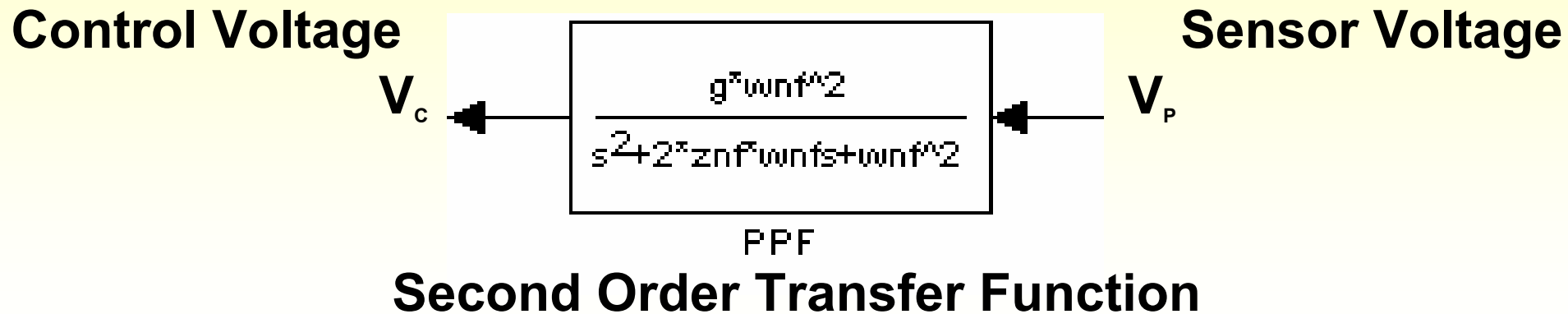
## 3. Experimental Verification



## 4. Conclusions



# A positive position feedback control loop was used for vibration reduction



Easy to use and very stable

Used displacement as control input

Behaves like tuned absorber

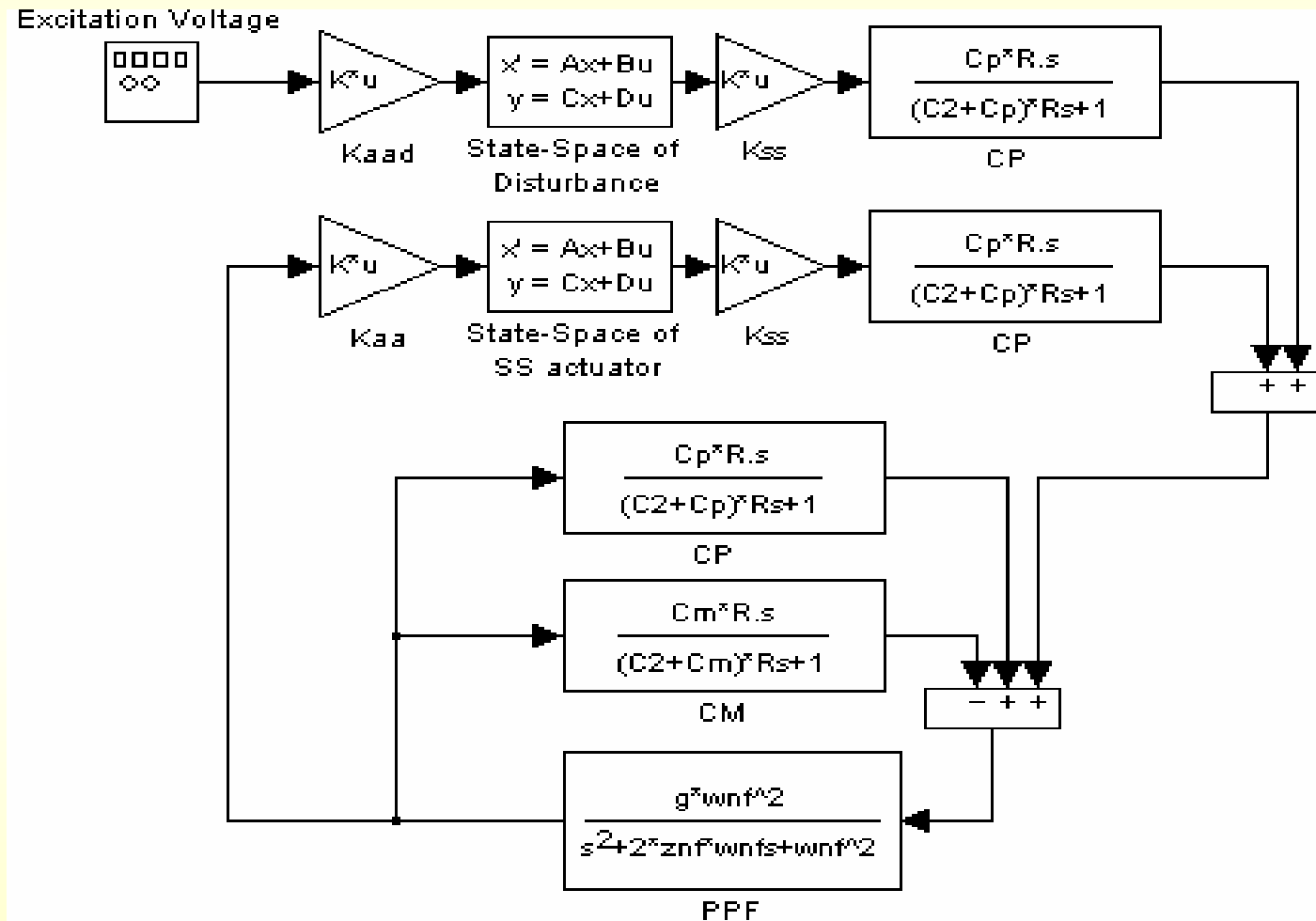
Natural frequency  $\omega_{nf}$

Damping  $\zeta_{nf}$

Gain  $g$  (moves pole further left in S-plane)

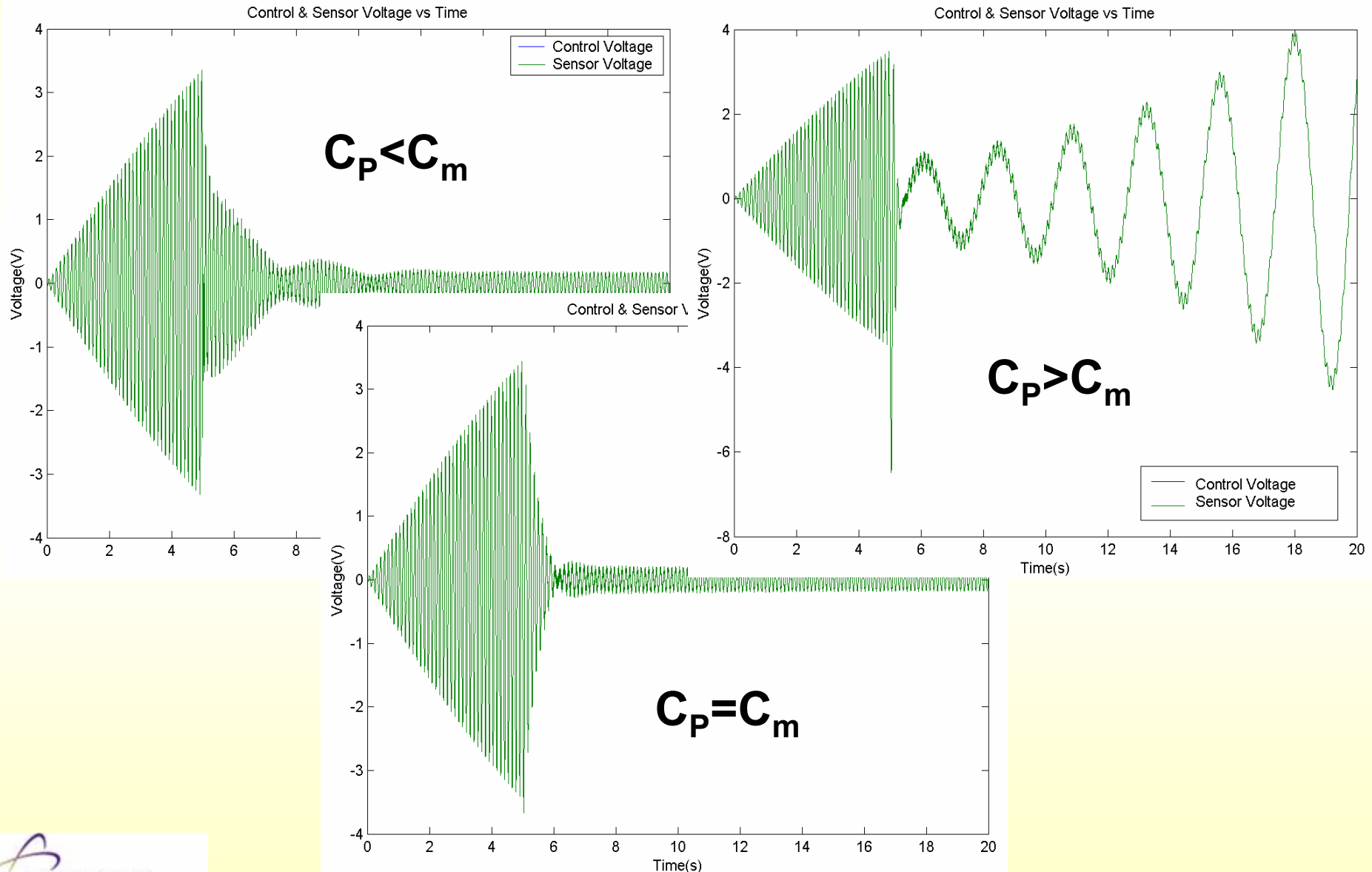
# The piezo-beam, self-sensing bridge, and feedback control were modeled analytically

With piezo-actuator for neutralizing and dynamic strain measurement



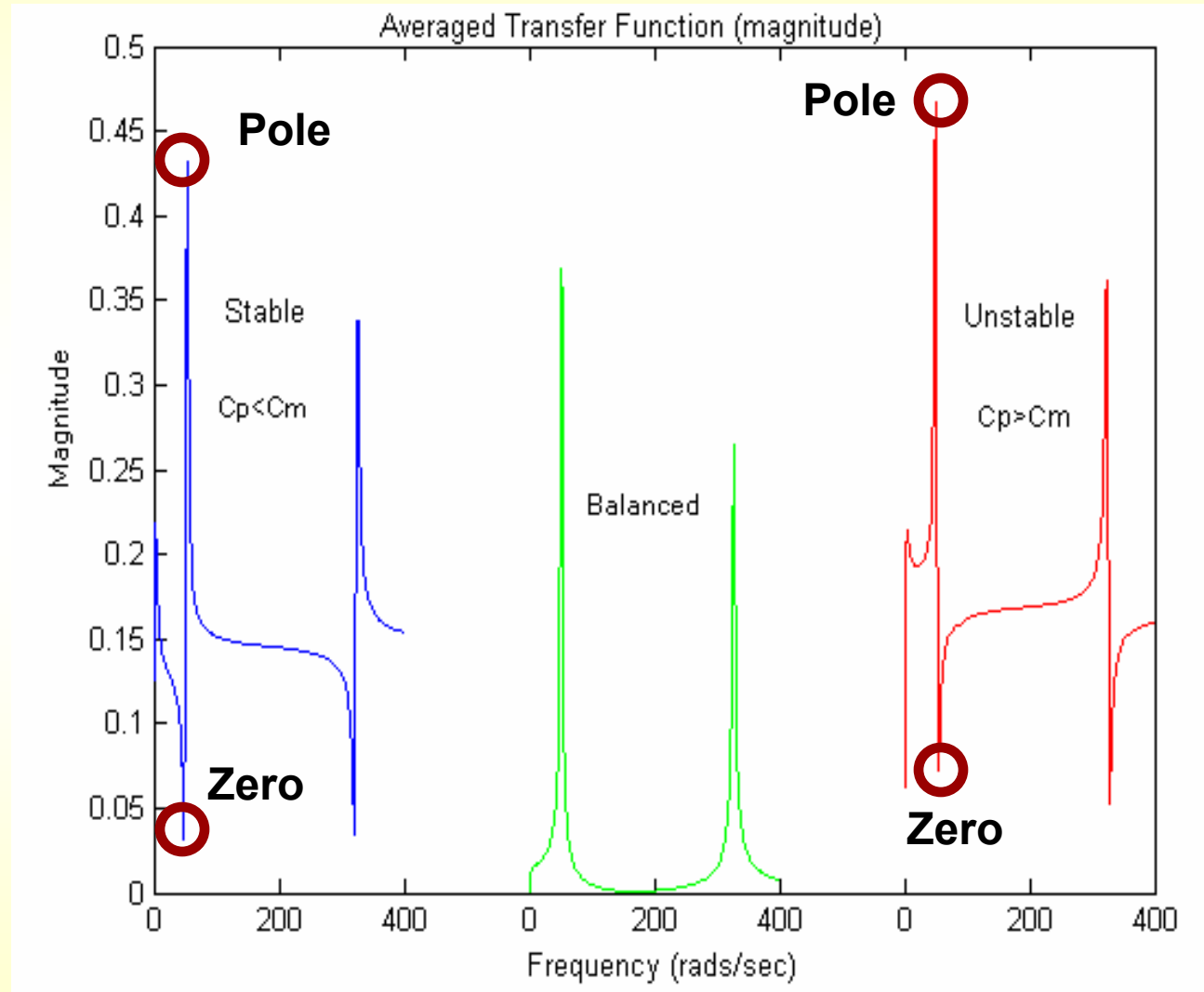


# A scenario was developed to identify how $C_p$ and $C_m$ related to stability

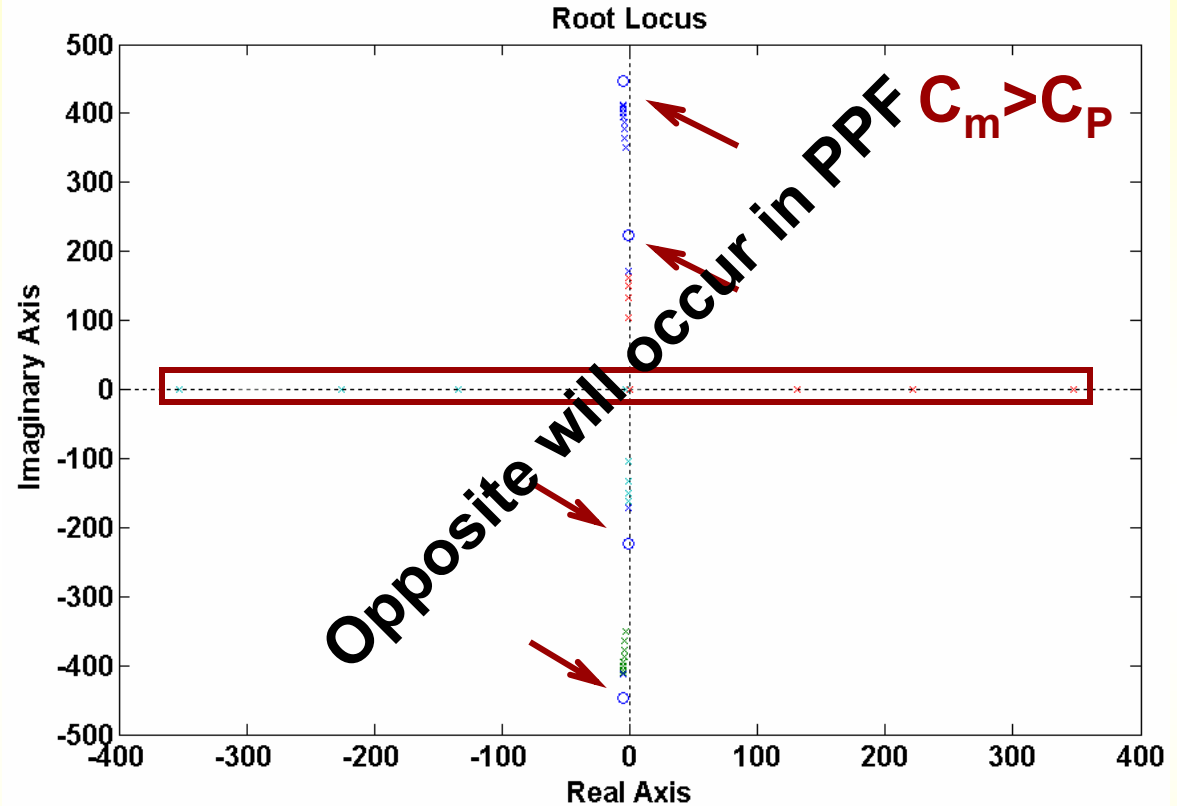
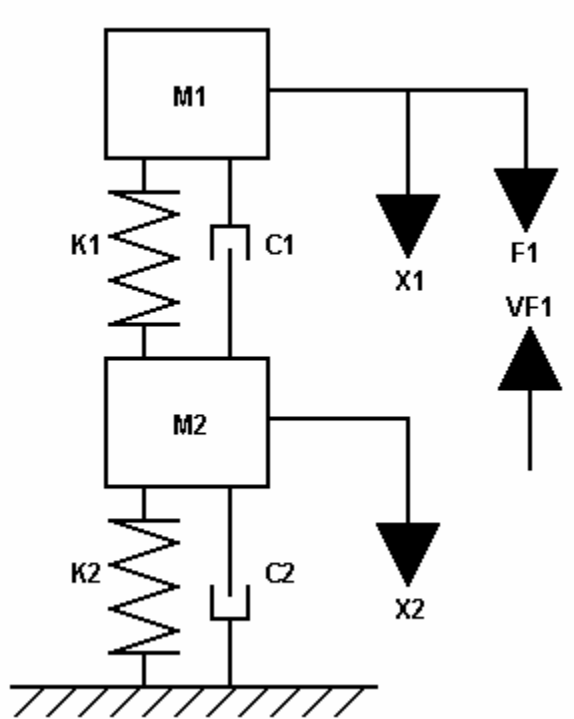


# What caused the system to become unstable?

## Analytical FRFs obtained via Simulink®



# The root loci of a 2 DOF model representing the self-sensing system was studied



$$\begin{aligned} X1 &= V_P \\ F1 &= V_C \\ VF1 &= V_m \\ V &= q/C \end{aligned}$$

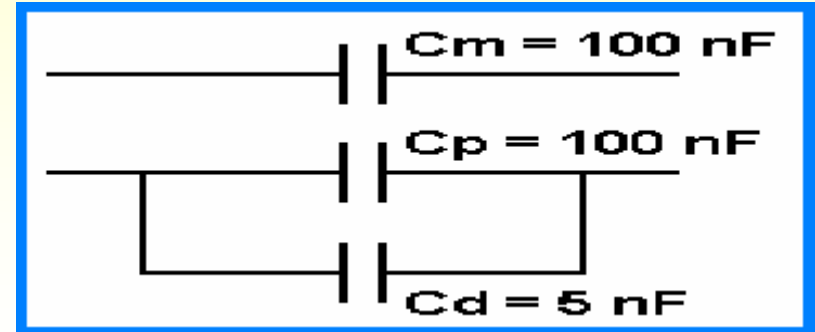
If  $C_m = C_P$ , then  $V_m = V_C$ , therefore  $X = X1$

If  $C_m < C_P$ , then  $V_m > V_C$ , therefore  $X = X1 - \alpha F1$

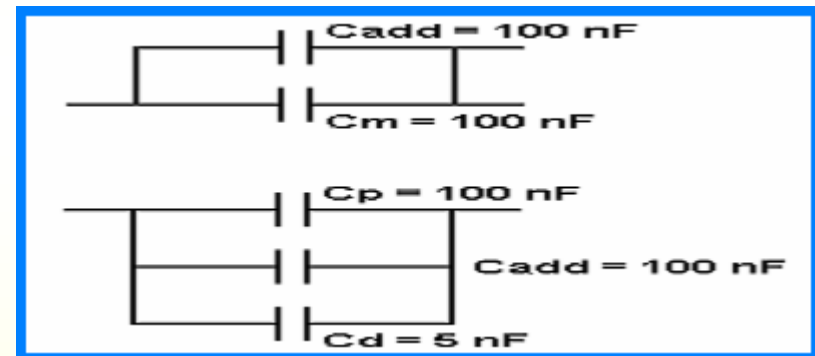
If  $C_m > C_P$ , then  $V_m < V_C$ , therefore  $X = X1 + \alpha F1$

# Our concept for improving the stability of the system was based on minimizing percent mismatch.

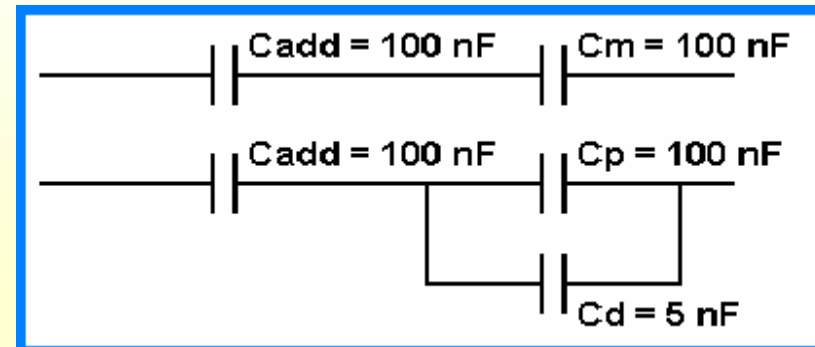
**No added capacitor case:  
5% mismatch**



**Added capacitor in parallel:  
2.5% mismatch**

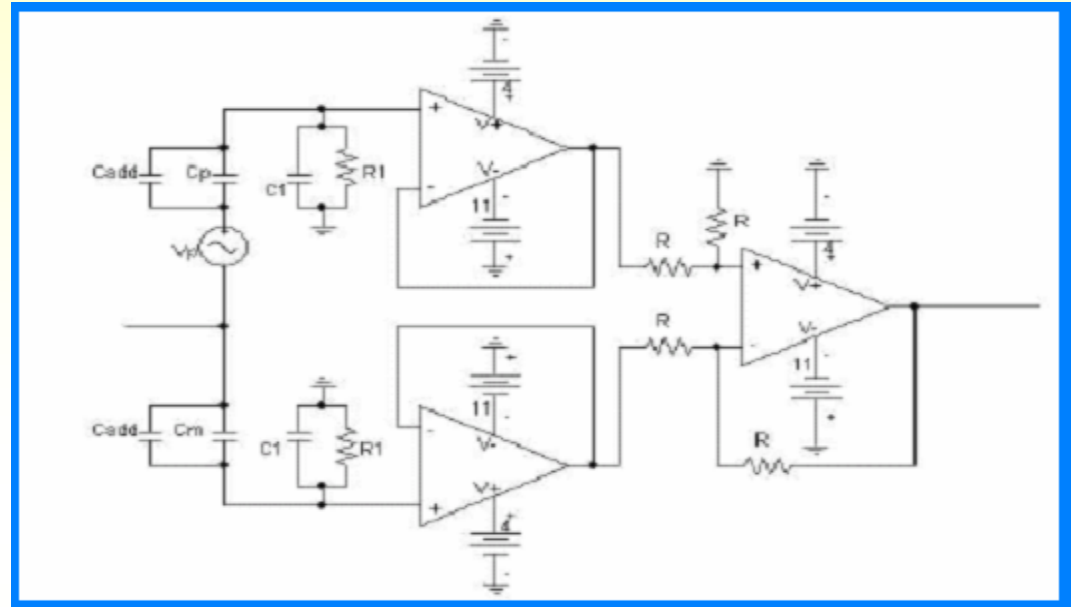


**Added capacitor is series:  
2.4% mismatch**



**Simulations were developed to test the added capacitor concept.**

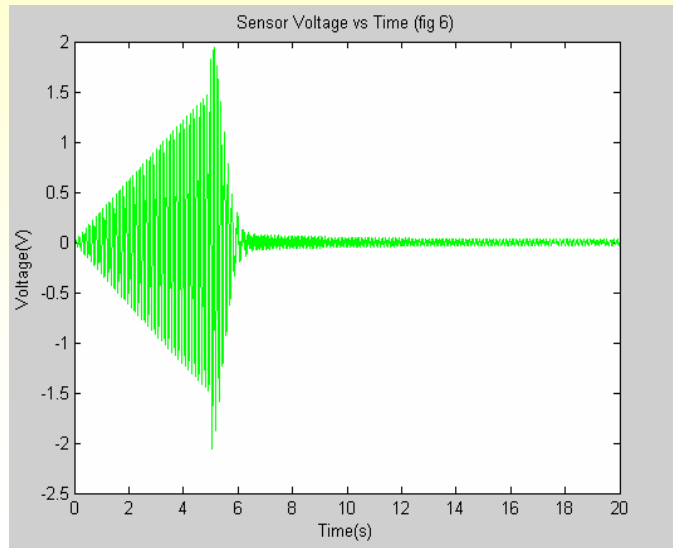
**Transfer functions were derived for the modified bridge circuits.**



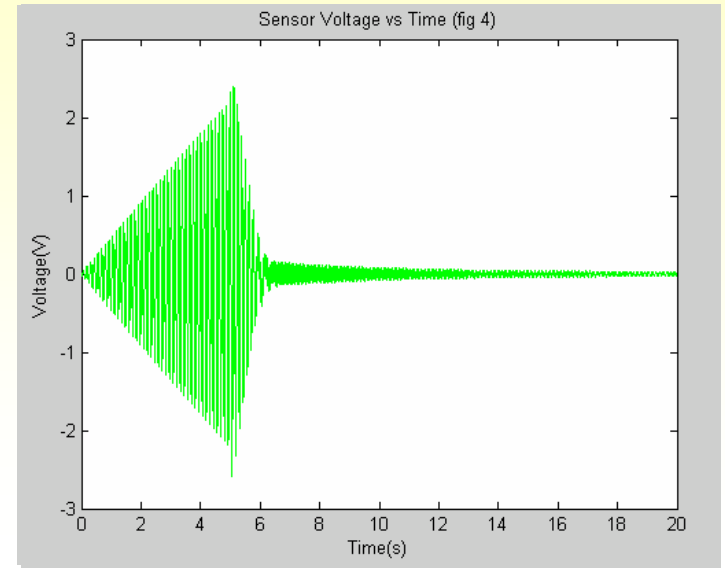
**The transfer functions were incorporated into the Simulink® model.**

$$V_s = \frac{(C_{add} + C_p) \cdot R_1 \cdot s}{(C_1 + C_{add} + C_p) R_1 \cdot s + 1} \cdot V_c - \frac{(C_{add} + C_{p21}) \cdot R_1 \cdot s}{(C_1 + C_{add} + C_{p21}) R_1 \cdot s + 1} \cdot V_c + \frac{(C_{add} + C_p) \cdot R_1 \cdot s}{(C_1 + C_{add} + C_p) R_1 \cdot s + 1} \cdot V_c$$

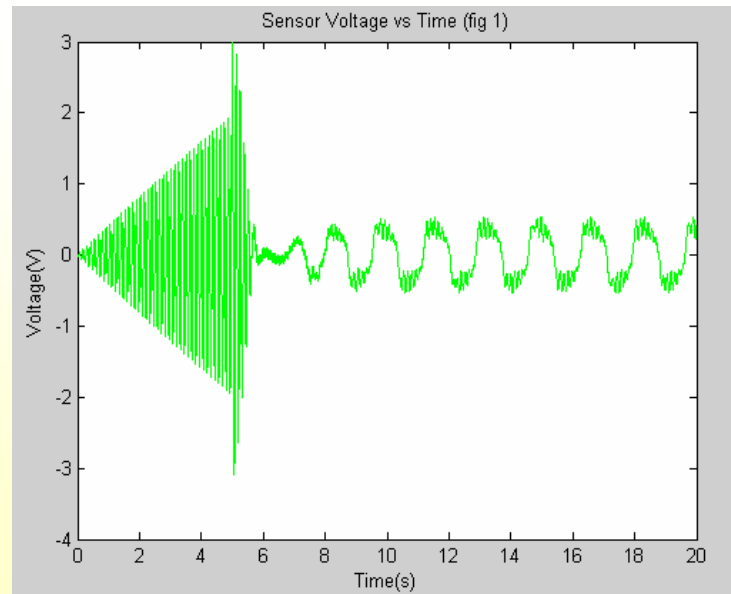
# A 6% temperature change: $C_{add}$ are stable



**Series add**  
**Stable**  
 $t_s = 1.4s$

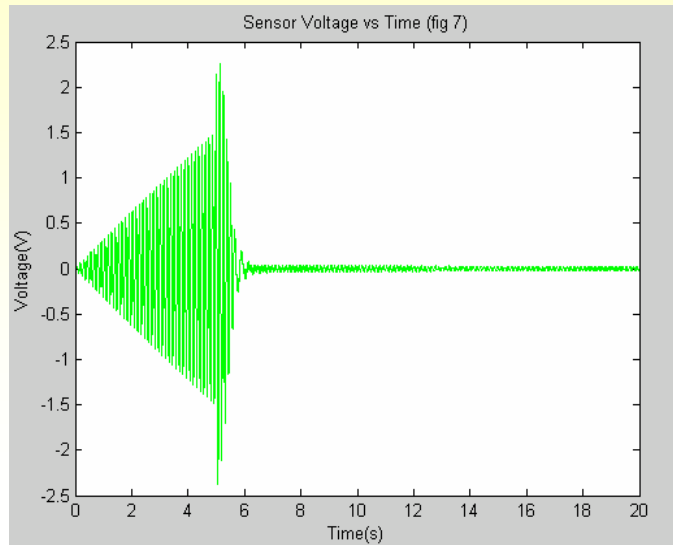


**Parallel add**  
**Stable**  
 $t_s = 1.4s$



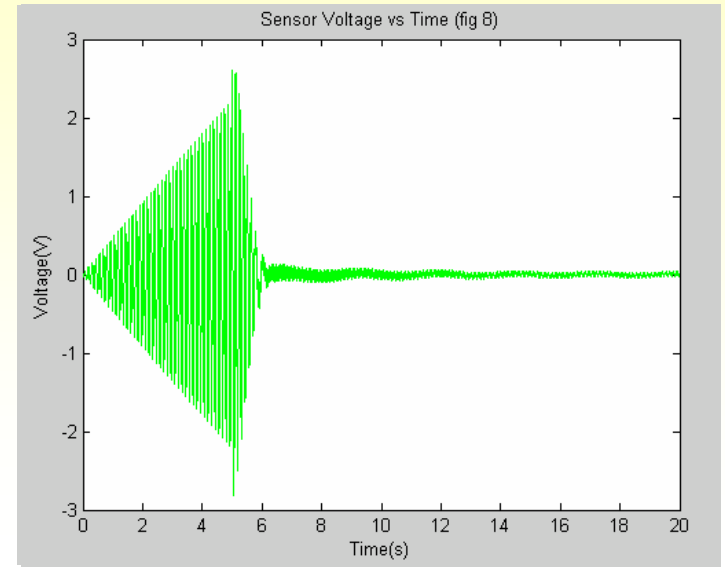
**No Add**  
**Unstable**

# At 9% temperature change: $C_{add}$ are stable

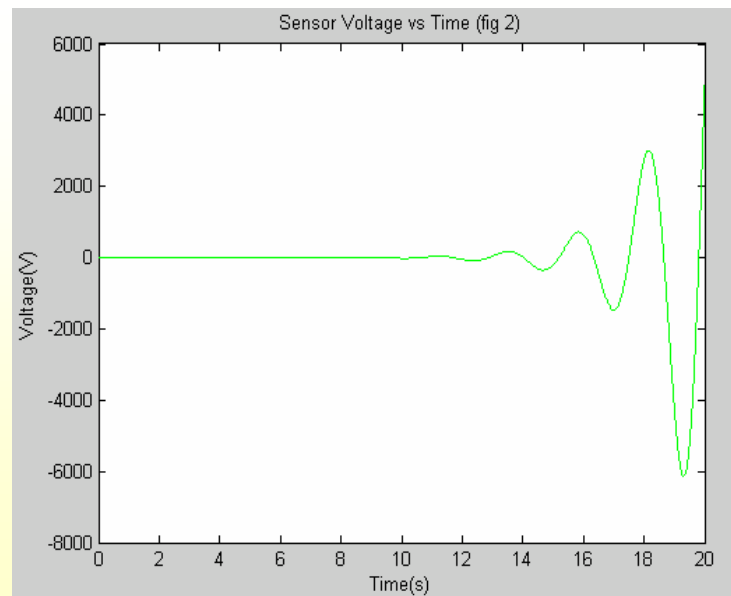


**Series add**  
**Stable**  
 $t_s = 1.4s$

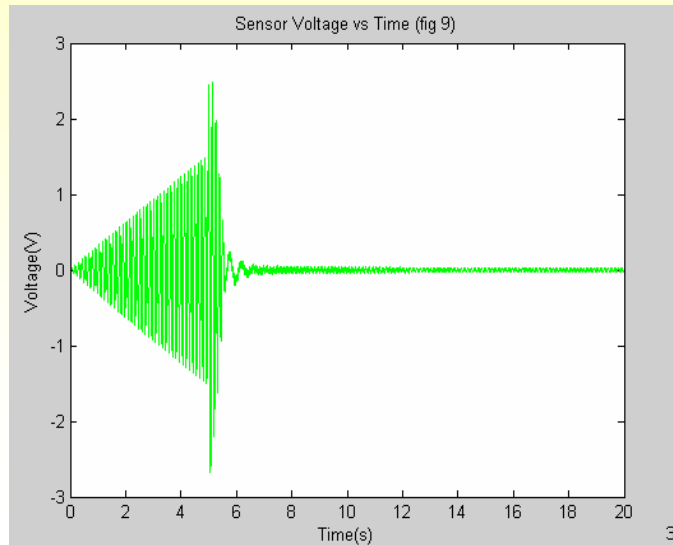
**No Add**  
**Unstable**



**Parallel add**  
**Stable**  
 $t_s = 1.4s$

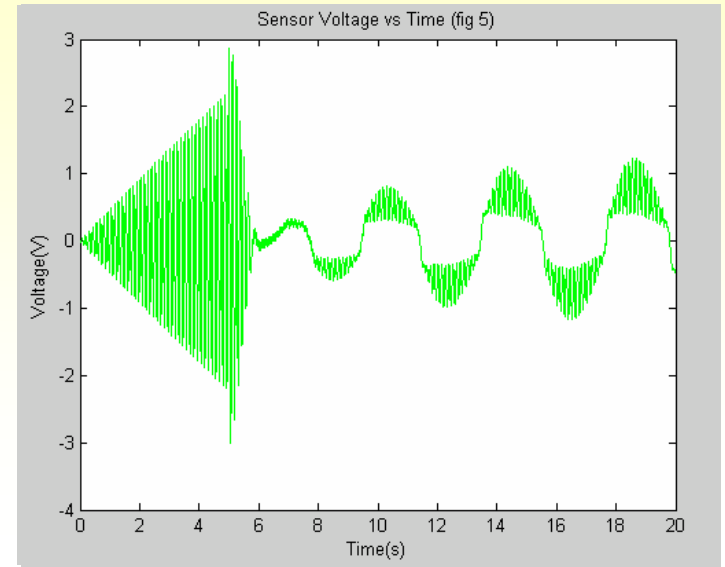


# At 12% temperature change: series is stable

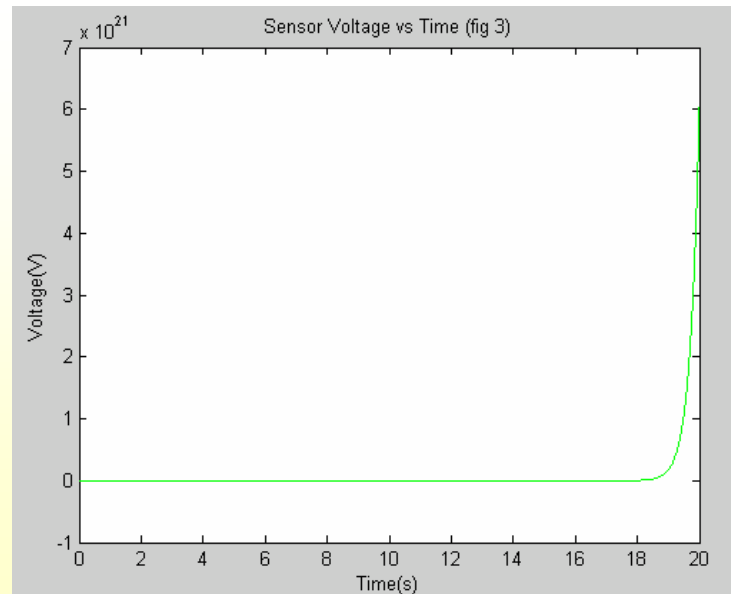


**Series add**  
**Stable**  
 $t_s = 2.21s$

**No Add**  
**Unstable**



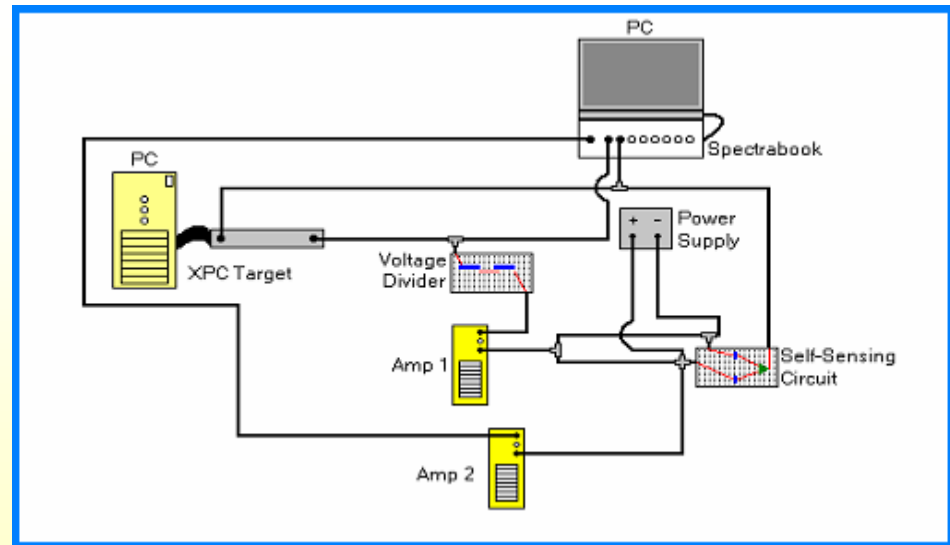
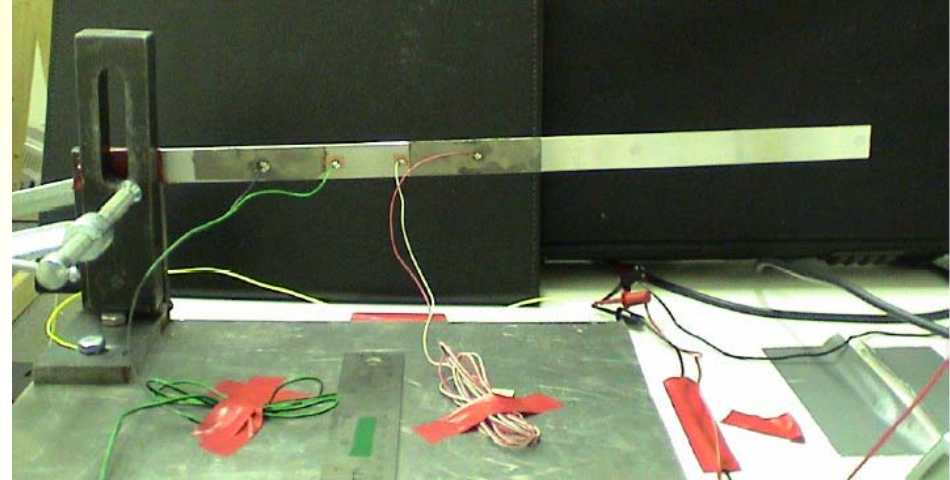
**Parallel add**  
**Unstable**





# The simulation results were verified experimentally by using an aluminum cantilever beam.

Length	0.398 m
Width	0.190 m
Thickness	0.00158 m
Root to Patch 1	0.0180 m
Length Patch 1	0.072 m
Between Patch	0.045 m
Length Patch 2	0.072 m
E	69 E 9
Base	0.335x0.311m
Base thickness	0.005 m



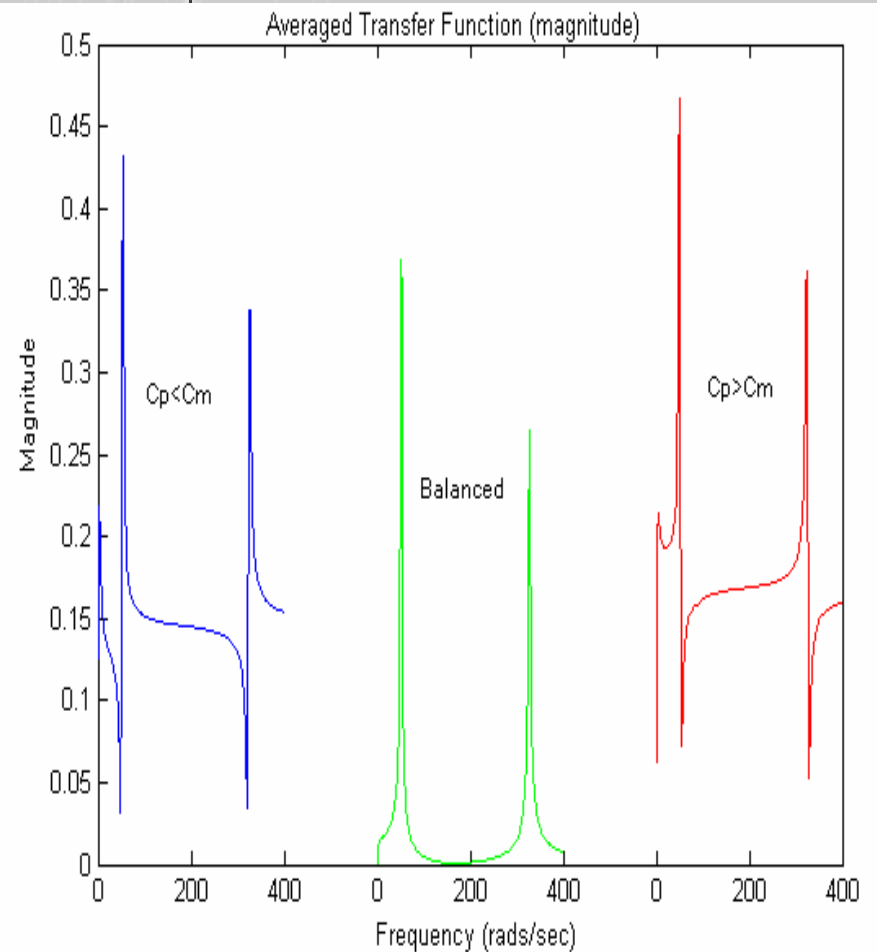
# Frequency response functions are related to system stability

When  $C_p = C_m$  the FRF shows no anti-resonance

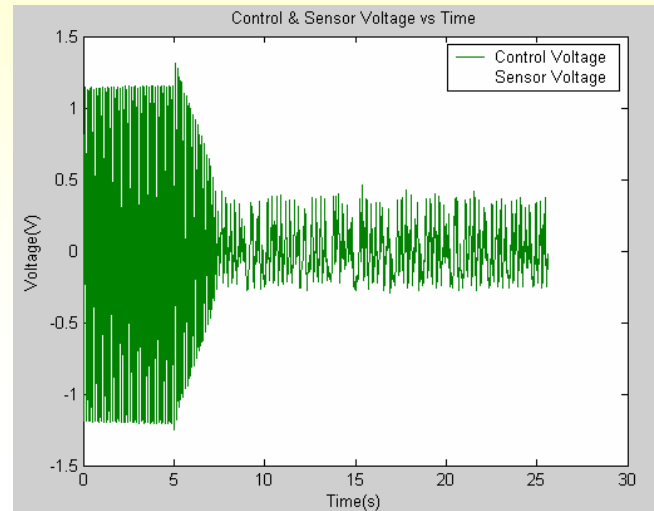
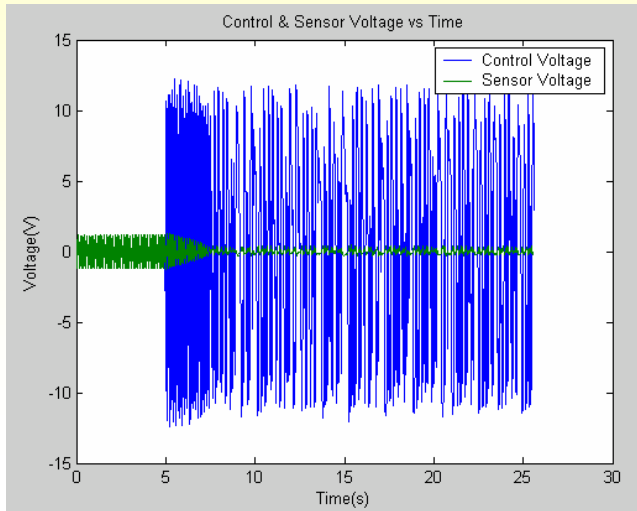
When  $C_p < C_m$  anti-resonance comes first

When  $C_p > C_m$  anti-resonance comes last

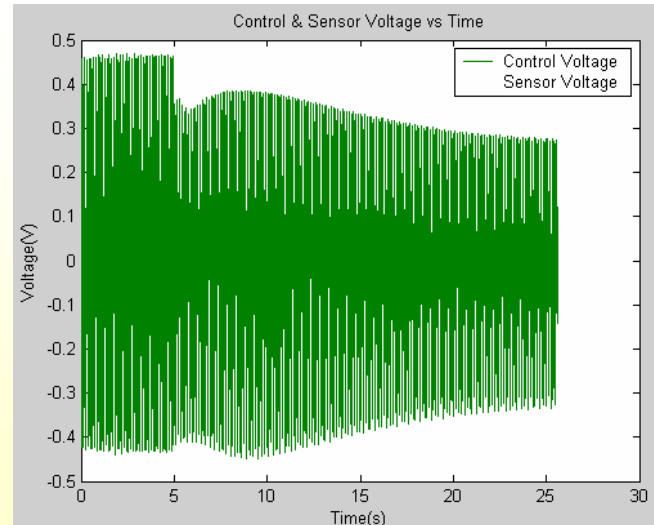
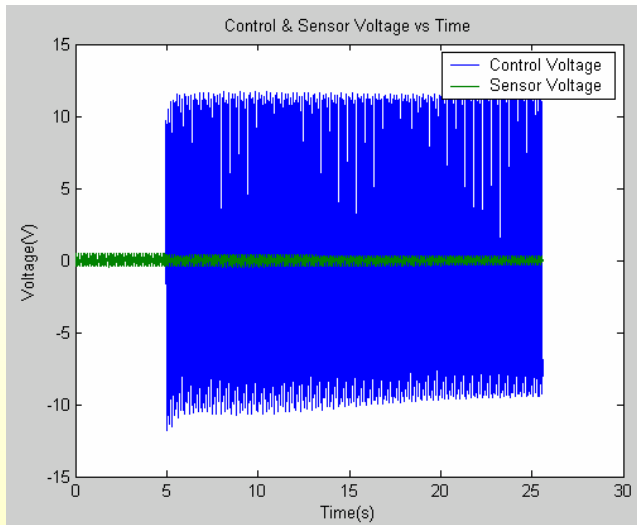
Changes in FRF due to  $C_p > C_m$  Analytical FRFs



# Below are some examples of experimental scenarios that are unstable or ineffective.

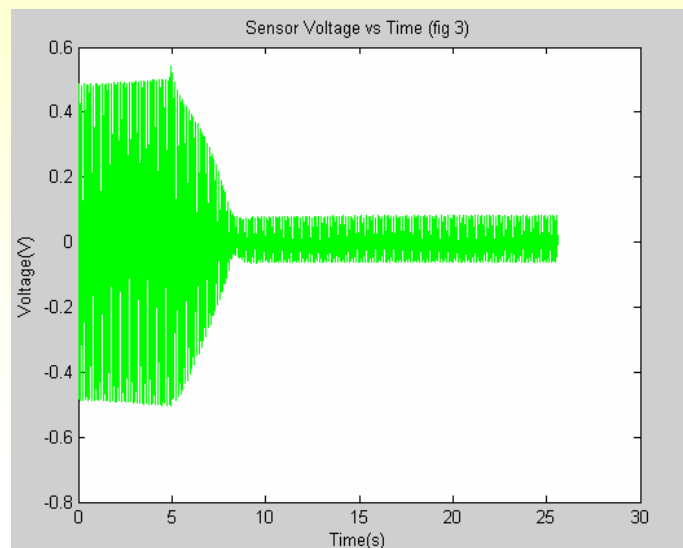


**Unstable**

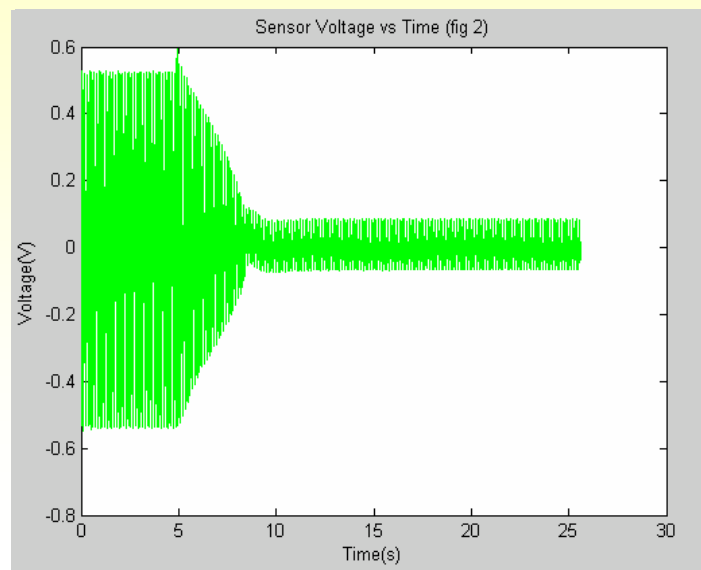


**Not Effective**

# With no temperature disturbance all cases are stable

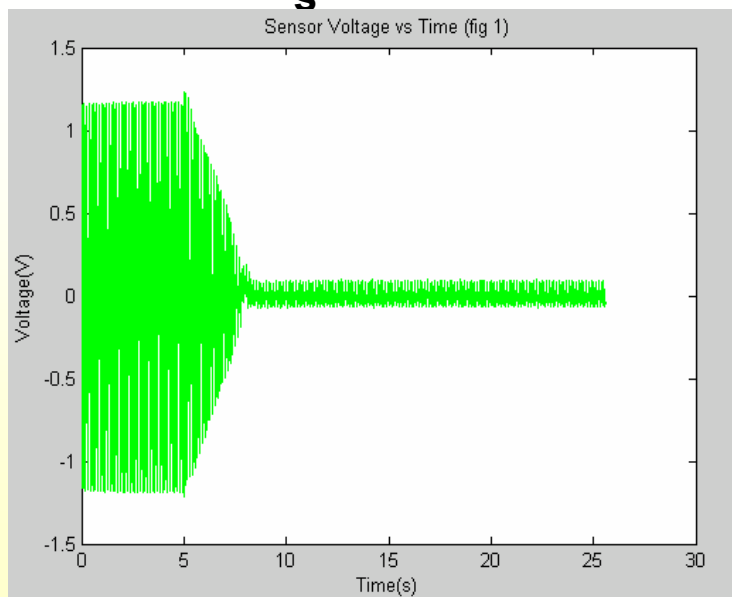


**Series add**  
**Stable**  
 $t_s = 3.35s$



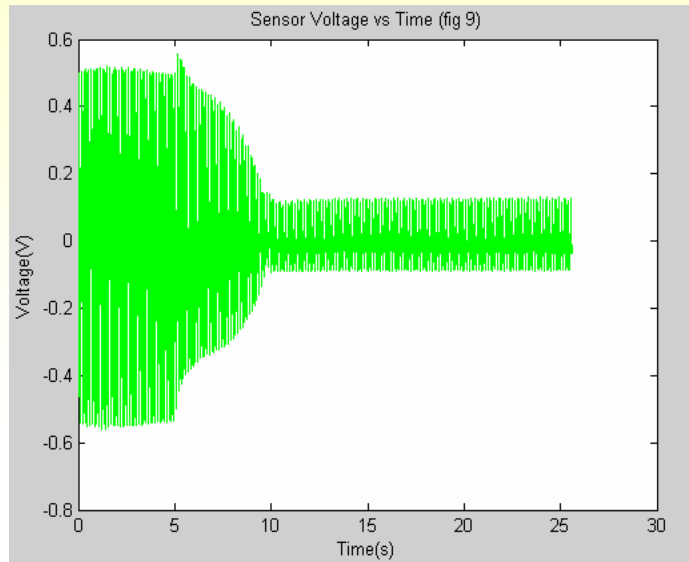
**No Add**  
**Stable**  
 $t_s = 2.81$

**Parallel add**  
**Stable**  
 $t_s = 3.46$



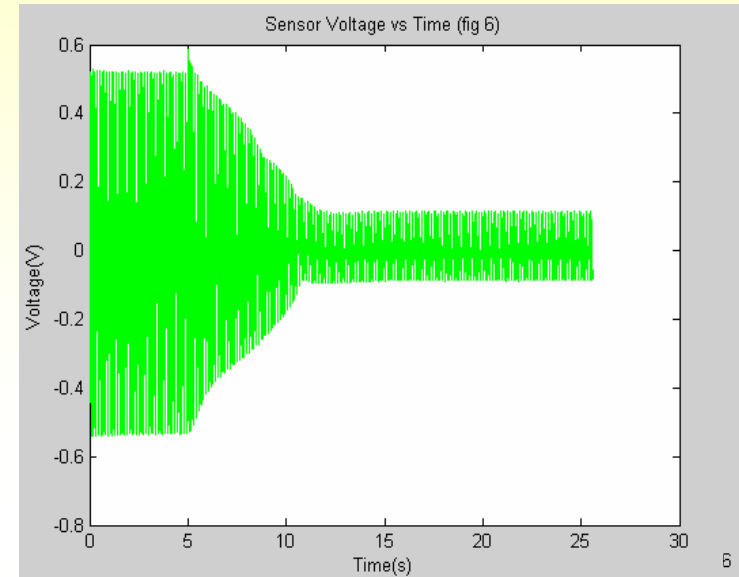
**20**

# For a 4 nF disturbance: $C_{add}$ creates stability

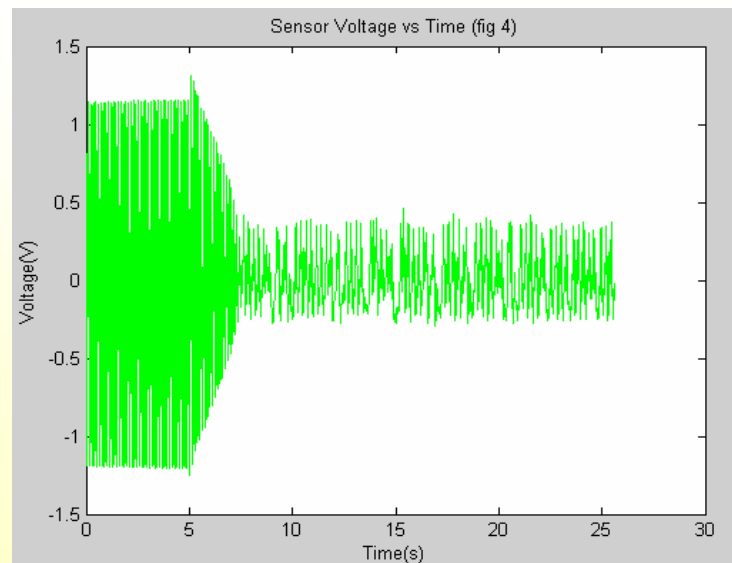


**Series add**  
**Stable**  
 $t_s = 5.03s$

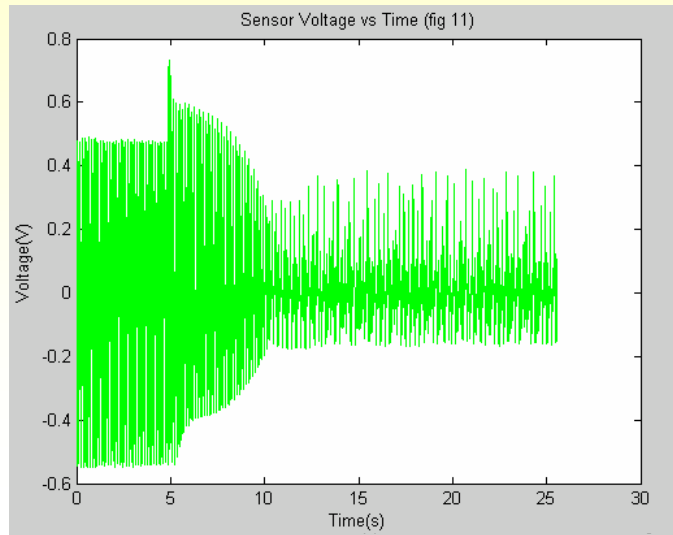
**No add**  
**Unstable**



**Parallel add**  
**Stable**  
 $t_s = 6.08s$

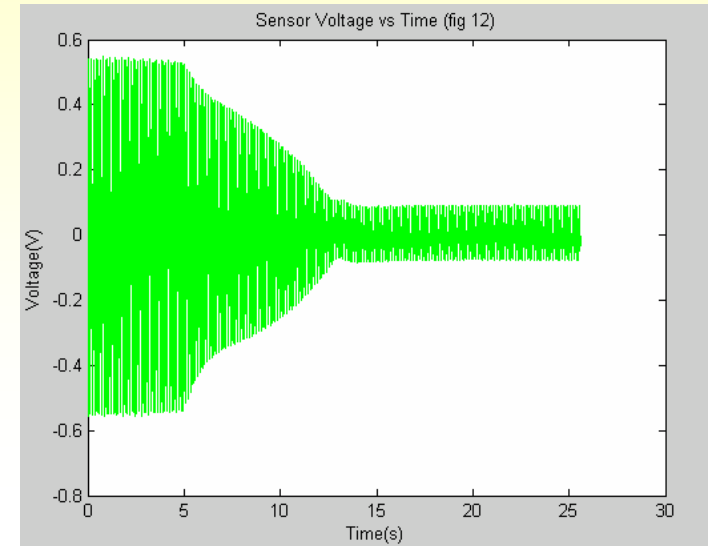


# 10 nF Disturbance: only the parallel case is stable

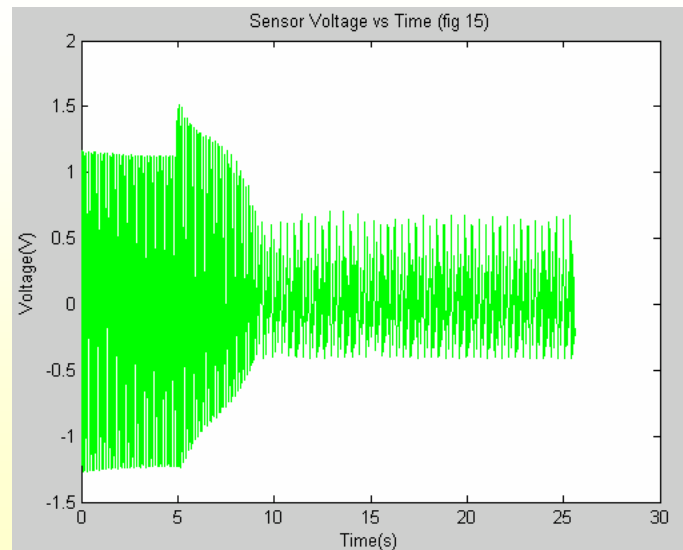


**Series add**  
**Unstable**

**No add**  
**Unstable**



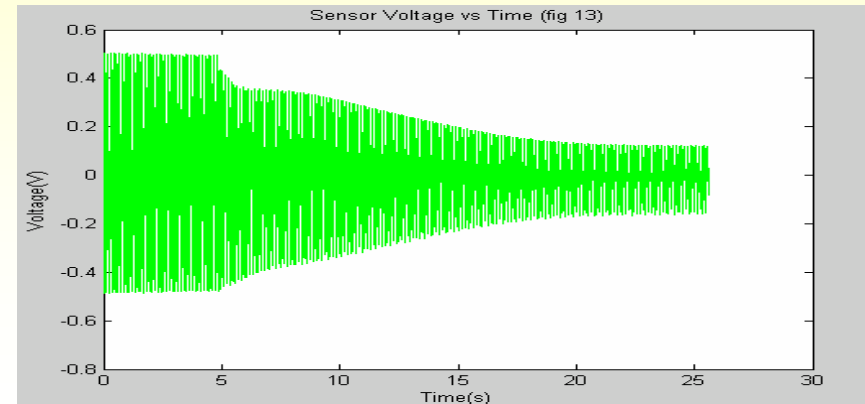
**Parallel add**  
**Stable**  
 $t_s = 7.70s$



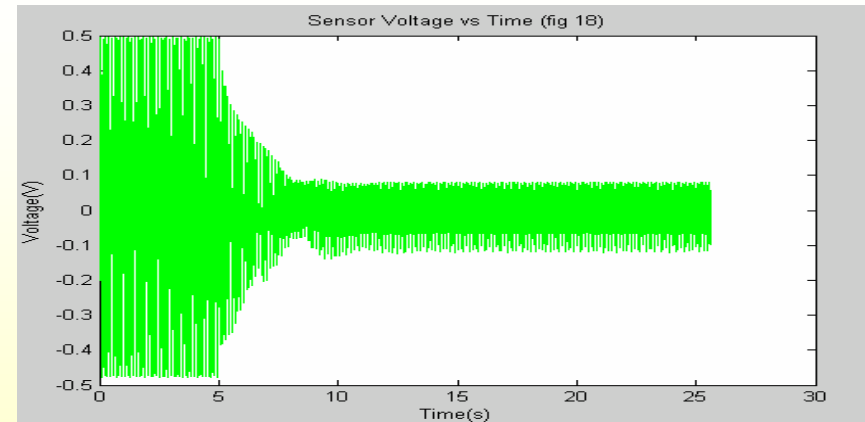
**Increasing the power to the amplifier makes the modified circuits more effective.**

**The loss of effectiveness is most prevalent when  $C_m > C_p$**

**Settling time can be decreased by increasing control power.**



**Low Power:  $t_s = 7.7s$**



**High Power:  $t_s = 3.0s$**

# Conclusions & Contributions

**Dynamic characteristics of the self-sensing actuation was quantified for the first time in literature.**

**Two new design schemes have increased control stability, which makes self-sensing more commercially viable.**

**The effectiveness of the two design schemes can be enhanced at the cost of increased power to the controller.**

**Both new design schemes were validated experimentally.**



# Recommendations

**What is the optimal value of the added capacitor?**

**Quantify the tradeoff between stability (temperature resistance) and effectiveness (vibration reduction).**

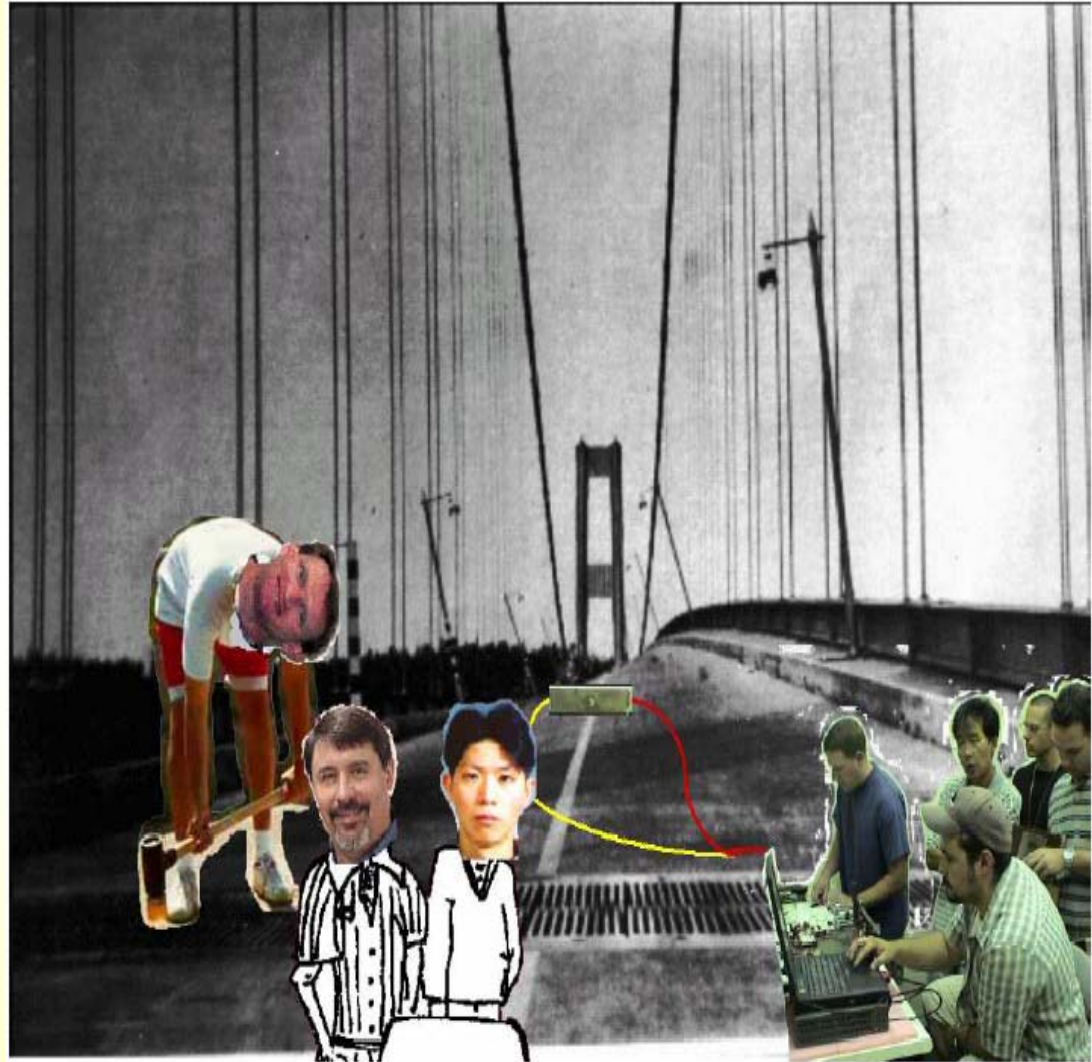
**Applying this technique to complex and real-scale structures.**

**Use this technique with damage detection schemes.**

# Acknowledgements

**We would like to thank:**  
**Dr. Gyuhae Park**  
**Dr. Hoon Sohn**  
**Mr. Henry Sodano**  
**Dr. Charles Farrar**  
**Dr. Pete Avitabile**

**Software Suppliers**  
**-MatLab®**  
**-Simulink®**

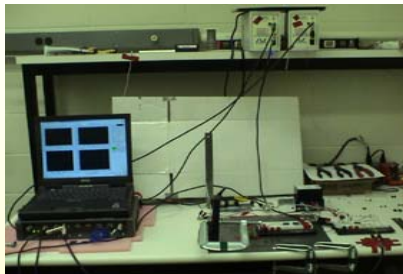
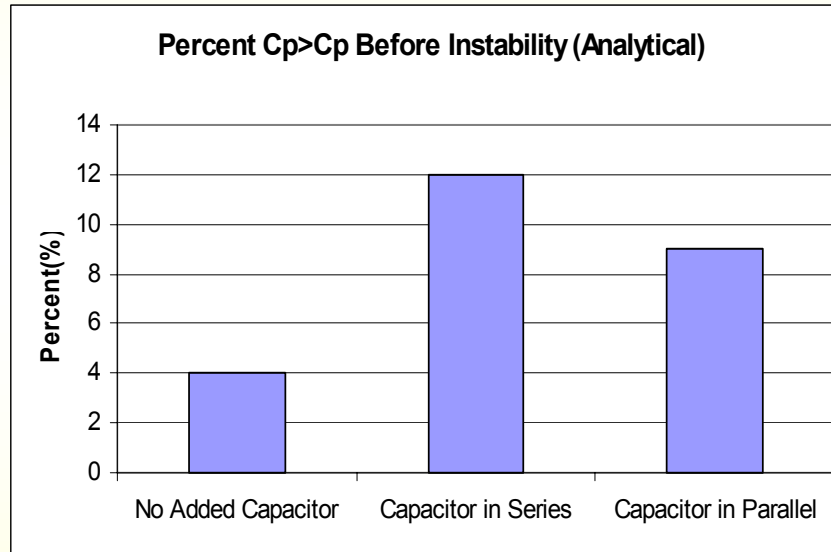


**The real DSS**

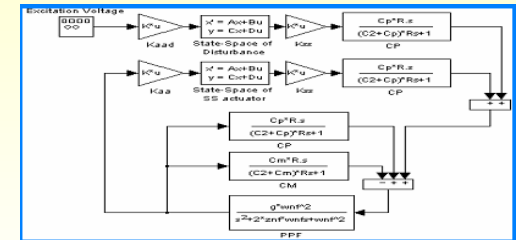
# Questions?

$$A = \begin{bmatrix} 0 & 0 & 0 & 0 & 1 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 1 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 0 & 1 \\ -\omega_1^2 & 0 & 0 & 0 & -2\zeta_1\omega_1 & 0 & 0 & 0 \\ 0 & -\omega_2^2 & 0 & 0 & 0 & -2\zeta_2\omega_2 & 0 & 0 \\ 0 & 0 & -\omega_3^2 & 0 & 0 & 0 & -2\zeta_3\omega_3 & 0 \\ 0 & 0 & 0 & -\omega_4^2 & 0 & 0 & 0 & -2\zeta_4\omega_4 \end{bmatrix}$$

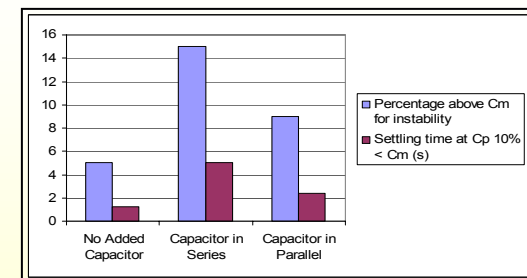
## Modeling



## Experimental



## Simulation



## Conclusions